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PAPR REDUCTION USING PTS, SLM AND CUCKOO SEARCH OPTIMIZED COMPANDING

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ABSTRACT

The OFDM is a multicarrier modulation technique has proved itself in the field of high-speed communication enabling current technologies to reach a high transmission rate. It is used in most communication standards. One of the major drawbacks of OFDM modulation is the strong PAPR of the resulting signals. This characteristic makes OFDM signals very sensitive to the nonlinearities of analog components, especially those of power amplifiers. Because for a high efficiency, the power amplifier must operate in a so-called non-linear (or saturation) zone, but it is in this zone that nonlinearities occur. The amplifier, sources of distortion of the signals to be transmitted. These effects are all the more annoying as the signals to be amplified are at high PAPR

This paper provides comparative analysis of several methods for PAPR reduction namely; Selective Mapping (SLM), Partial Transmit Sequences (PTS) and Companding Transform optimized by Cuckoo Search (CS) algorithm. Simulation results are presented by using CCDF v/s PAPR graph.

Keywords –BER, CCDF, Companding, Cuckoo Search, HPA, OFDM, PAPR, PTS, SDR, SLM.

INTRODUCTION

Telecommunications systems, and more particularly wireless communication systems have evolved dramatically over the past two decades. This evolution went hand in hand with the achievements made in digital electronics. Wireless communication systems have become almost ubiquitous in our lives. In our pockets, we can find telephones, diaries, satellite positioning systems, all communicating with each other and with the outside. In the office or at home, we are surrounded by data storage devices, computers, multimedia and television centers, alarms and sensors, which also communicate. And all these wireless communications, involving freedom, availability and comfort, can be established through the transfer of information over the airwaves [1].

Despite this spectacular evolution in the field of telecommunications, new challenges are facing manufacturers today [1]. Given the socio-economic factors, the demand of the users in term of communication is precise: to transmit and / or to access a volume of information (of more and more important size), as fast as possible, with the greatest possible flexibility, all with limitless mobility and miniaturized media. Nowadays, we are no longer content to speak with a contact person on the phone, but we demand a good quality of the communication and often we want to see the image in real time of our interlocutor on our communication support (video telephony, teleconference, etc.). Operators of telephony, too, want to maximize the number of subscribers, the quality that is available to them, and the number of calls that are established, and therefore wish to maximize the quantity of information and the capacity of networks. Increasingly greedy users of data, information and capacity on one side; and competing providers to best meet the needs in terms of the quality and quantity of user information.

To respond to this multitude of requests, we have also witnessed a proliferation of communications standards for new types of modulations, and so on. Often, these communications standards have radio parts incompatible with each other. And to answer the problems of incompatibility between communication standards, the concept of Software Defined Radio (SDR) has been introduced [2].

Software Radio implies a multi-standard, multi-modulation, multi-mode, re-configurable radio link context. It is a technology capable of modulating or demodulating any signal anytime, anywhere. This implies a great diversity of bands of frequencies to be treated, modulation techniques and different access to take into consideration.

Before reaching the concept of Software Radio, the OFDM [3] is a sophisticated multi-carrier modulation technique which has proven itself in the field of wireless communication. Current technologies to achieve a high rate of transmission. It is used in most communication standards. In addition, ADSL (Asymmetric Digital Subscriber Line) is a communication technique that makes it possible to use a subscriber line to transmit and receive digital signals. It has high debit rates (high-speed internet), WI-FI, WIMAX (belonging to the IEEE 802.xx standard family and allowing transmission of high-speed wireless data [4, 5].

These technologies based on orthogonal multicarrier modulations have advantages as well as disadvantages. Advantages mainly relate to the robustness of the signal with respect to the multitrajet channel with fading and optimal spectralspace.

One of the fundamental disadvantages is that OFDM signals have a strong envelope fluctuation which is characterized by a high "crest factor". The "crest factor" is equivalent to PAPR [6].

This high PAPR makes the OFDM signals very sensitive to the non-linearity of the analog components, especially those of the power amplifier.

The high power amplifier is a decisive element in a communication chain insofar as it has a preponderant influence on the overall transmission balance in terms of power, efficiency and distortion. In terms of consumption, the power amplifier is the element that consumes the most energy among all elements of the issuer. It is therefore important to make it work as efficiently as possible, especially for mobile devices where consumption is a decisive factor in determining their autonomy. Unfortunately, for a high efficiency, the power amplifier must operate in a so-called nonlinear (or saturation) zone, or it is in this zone that the non-linearities are presented. The effects of the amplifier, sources of distortion (inter-modulation, spectral rise, etc.) of the signals to be transmitted. These effects are all the more annoying as the signals to be amplified are at high PAPR. Often, it is necessary to take a sufficient step back so that the power amplifier does not saturate the signal to be amplified. This amounts to making the power amplifier work in a highly linear zone with a significant deterioration of its efficiency and therefore with an increase in the overall consumption of the mobile terminal for the same coverage area.

It is necessary to optimize the power amplifier's consumption by making it operate as close as possible to its saturation zone, which represents the zone at optimum efficiency. In order to avoid itself for the saturation zone without saturating the input signal too much, it is necessary to decrease the envelope fluctuations of the OFDM signal, and thus its PAPR. The first extensive studies on the reduction of the PAPR appeared concomitantly with the OFDM modality-based communication standards in the early 1990s. A multitude of techniques for reducing PAPR. Given the large number of these methods, the reader may refer to [7] or a general classification of methods of reducing PAPR and correcting nonlinearities are proposed. Some techniques act on the amplifier to avoid saturation of the input signal and other techniques are based on a processing performed directly at the signal level.

However, acting on a signal to reduce its PAPR is frequently to the detriment of expanding the involvedness of the system and / or the escalation in average power, bit error rate (BER) and / or the secondary lobe rise and / or the decrease of the useful flow. However, it seems that signaling techniques are being watched more and more closely, particularly because of their strong capacity to reduce the PAPR, and often because they do not degrade the BER and are generally downwardly compatible. A "signal addition" technique is a technique that allows the reduction of the PAPR by adding one or more signals called additional signals or PAPR reduction signals or correctional signals [8]. These are particularly attractive and promising PAPR reduction techniques; they have also been standardized in DVB-T2.

This paper proposes a three techniques for PAPR reduction:

- Selective Mapping
- Partial Transmit Sequences
- Cuckoo Search Optimized Companding

Section two presents an overview for various techniques or PAPR reduction. Section three describes the performance criteria for PAPR reduction. Section four presents the proposed methodology used for research work followed by the simulation results and conclusion in section five and six respectively.

STATE OF THE ART OF PAPR REDUCTION TECHNIQUES

As we have seen in the previous heading, in order to get the amplifier the linearity required by the communication standards, the power amplifier can be oversized which amounts to working away from the saturation zone, with a significant decline. On the other hand, the energy efficiency in this case will prove very low, and this, all the more so as the waveforms of the signals will have high PAPRs. However, a high PAPR signal passing through a nonlinear amplifier (with a low recoil, this for a high efficiency) becomes much distorted. Compensation for this distortion in reception involves complex processes. This is why most of the nonlinearity processing techniques proposed in the literature are on the program.

The purpose of this heading is to establish a state-of-the-art of techniques for reducing the PAPR to the program knowing that in this topic we will be interested in more the techniques for reducing the PAPR to the so-called "signal addition" program because they are simple to implement and satisfy the constraints (resource consumption) of the on-board systems. In addition, some of them are downwardly compatible. We will return to the notion of top-down compatibility, which in our view is very fundamental in our analysis and research of the technique of reducing the PAPR to the program.

First of all, it should be noted that the problem of reducing the PAPR has come about at the same time as the problems of the PAPR analysis. By the end of the 1950s, Shapiro [9] and Rudin [10] focused on reducing the PAPR of signals by finding adapted sequences. Schroeder in [11] has instead focused on the generation of signals with low PAPR. However, it is only with the popularization of the OFDM modulation (due to its use in many telecommunication standards such as DVB-T, ADSL, DAB or in the IEEE standard 802.11a / g) that the problem

has become more crucial because the signal has a non-constant envelope. Thus we have seen the first techniques of treatment of nonlinearities from the 1990s.

In [7], a general classification of all the techniques dealing with the problem of non-linear amplification has been proposed, this classification has been repeated and put into practice. Here, in this research work, we propose a classification of techniques for reducing PAPR of OFDM system. This is an extension of the classification proposed by Y. Louet and J. Palicot [7] in which all distortion techniques (Distortion techniques means any reduction technique of PAPR that generates distortions, as is the case in clipping techniques) will be considered as adding techniques of signal.

There is further work in the classification of PAPR reduction techniques. We can cite, among others, the work of [12] where they presented OFDM PAPR reduction techniques by proposing a SLM based technique which is based on computational complexity, bandwidth expansion, transmission rate decrease and the variation of the average signal power to be transmitted.

PERFORMANCE CRITERIA FOR PAPR REDUCTION TECHNIQUES

This section defines the different criteria of equity that will allow us to evaluate the performance of a particular program. The technique of reducing PAPR in its operating environment. Consider the simplified scheme of a transmission chain (Figure 1) incorporating a reduction module of the PAPR.

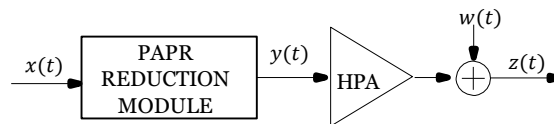


Figure 1: Simplified scheme of a transmission chain incorporating a PAPR reduction module

A. Performance in Reducing the PAPR

This may seem trivial but it remains despite all the determining criteria. To illustrate the measure of performance in reducing the PAPR of a method, consider Figure 1. The PAPR of the signals $x(t)$ and $y(t)$ are defined as follows:

$$\begin{aligned}
 PAPR_{[x]} &= \frac{\max_{0 \leq t \leq T_s} |x(t)|^2}{\mathcal{P}_x} \\
 PAPR_{[y]} &= \frac{\max_{0 \leq t \leq T_s} |y(t)|^2}{\mathcal{P}_y}
 \end{aligned}
 \tag{1}$$

Where T_s denotes the duration of an OFDM symbol and \mathcal{P}_x and \mathcal{P}_y , respectively represent the average powers of the signals $x(t)$ and $y(t)$. The PAPR is a random variable that can be characterized by its CCDF; $PAPR_{[x]}$ and $PAPR_{[y]}$ defined by equation (1) are random variables whose CCDFs are shown in Figure 2. The gain in the reduction of the PAPR is defined at a given ϕ value of the CCDF, the parameter $\Delta PAPR(\phi)$ defined by:

$$\Delta PAPR(\phi) = PAPR_{[x]}(\phi) - PAPR_{[y]}(\phi)
 \tag{2}$$

Where $PAPR_{[x]}(\phi)$ and $PAPR_{[y]}(\phi)$ are the PAPRs of the signals $x(t)$ and $y(t)$ to $CCDF = \phi$.

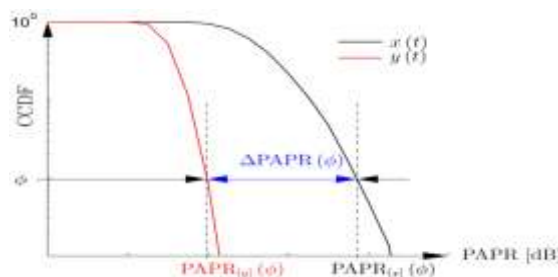


Figure 2: Reduction of the PAPR at a particular value of the CCDF [13]

B. Variation in Average Power

Consider the simplified scheme of the transmission chain given in Figure 1, the variation of the mean power of the signal to be transmitted is written as [14, 15]:

$$\Delta E[\tau] = \mathcal{P}_y - \mathcal{P}_x \text{ [dB]}
 \tag{3}$$

Where τ is the parameter that controls the reduction performance of the PAPR. The parameter τ can be assimilated to the "cut-off threshold" in the case of "clipping" [15].

C. Degradation of the BER

For some techniques, the reduction of PAPR is accompanied by the generation of distortions both outside and inside the useful signal band; this is the case of distortion techniques [16, 17, 18, 19, 20]. The reduction of the PAPR is not the only cause of degradation of the BER; indeed, the power amplifier can be a source of degradation of the BER inasmuch as by the need of energy efficiency, the amplification is in (or near) the saturation zone. On the other hand, the total degradation is defined by:

$$TD[\tau, IBO] = \Delta_{E_b/N_0}[\tau, IBO] + IBO, [\text{dB}] \quad (4)$$

Where IBO=Input Back-Off. And $\Delta_{E_b/N_0}[\tau, IBO]$ is the reduction of SNR for a given BER. In [71], the authors have decomposed this degradation of the signal-to-noise ratio into two components: a degradation due to non-linear amplification and a degradation due to the reduction of the PAPR if the technique used is a distortion technique.

$$\Delta_{E_b/N_0}[\tau, IBO] = \Delta_{E_b/N_0}[IBO] + \Delta_{E_b/N_0}[\tau], [\text{dB}] \quad (5)$$

Where $\Delta_{E_b/N_0}[IBO]$ is the degradation of the signal-to-noise ratio due to non-linear amplification and $\Delta_{E_b/N_0}[\tau]$ is the reduction of SNR due to the reduction of the PAPR. For a non-distorted PAPR reduction technique as is the case for Tone Reservation, there is no degradation of the SNR due to the reduction of the PAPR, i.e., $\Delta_{E_b/N_0}[\tau] = 0$.

D. Overall Performance

There are studies in the literature [22, 23, 24, 25, 26, 27, 28, 29, 30, 31, 32 and 33] on the performance of PAPR reduction techniques. In general, these studies take into account the reduction of the PAPR, the quality of signal transmission (that is to say the BER reduction and the increase of the Adjacent Channel Power Ratio (ACPR)) and the complexity of the technique. Unfortunately, there is very little study of the overall performance of a PAPR reduction technique in a more "realistic" environment, that is, in the occurrence of a non-linear type power amplifier and an AWGN type propagation channel, for example.

In [34], authors have proposed an overall performance (OP) metric which allows to evaluate the overall performance of a system and which takes into account the reduction of the PAPR of the system. The average power variation of the signal to be transmitted due to the reduction of the PAPR and the energy degradation per bit of the system. Since the average power variation and the energy degradation per bit of the system have efficient effects on the generalized performance of the system and only the reduction in the PAPR contributes positively to the performance of the system, we define the metric "OP" by the following relation:

$$OP[\tau, IBO] = \Delta PAPR[\tau] - \Delta E[\tau] - \Delta_{E_b/N_0}[\tau, IBO] \quad (6)$$

Where $\Delta PAPR[\tau]$ is the reduction gain of the PAPR, $\Delta E[\tau]$ is the variation of the mean power of the signal to be transmitted given by the equation (3) and $\Delta_{E_b/N_0}[\tau, IBO]$ is the energy degradation per bit given by the equation (5).

In the manifestation of $OP[\tau, IBO]$, one can incorporate the complexity of the system's PAPR reduction module and the out-of-band power emissions (which are the cause of the increase of the ACPR).

E. Complexity

If a technique is so proficient for the PAPR reduction but requires a lot of computation resources, this can become prohibitive for some practical applications such as "real-time". The complexity of the algorithms used must then be studied.

PROPOSED METHOD

F. Selective Mapping (SLM)

The idea is to multiply the sequence of complex symbols resulting from numerical modulation by a series of L different vectors so that only the product with the PAPR is retained lowest (after IFFT). This method does, however, require the transmission of redundancy information for the receiver to identify the optimal vector. This technique was presented by R. Bauml, R. Fischer and J. Huber [35]. It was then more detailed in [36, 37] by S. Muller and J. Huber.

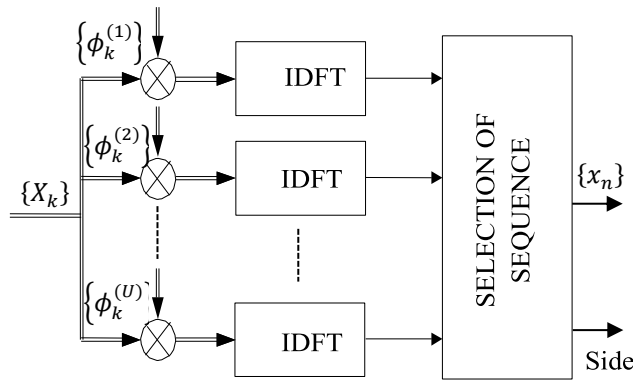


Figure 3: Selective Mapping scheme for PAPR reduction

This method applies to OFDM for any number of subcarriers and for any digital modulation. Let $X = \{X_k\}, k = 0, \dots, N - 1$, the OFDM symbol vector in the frequency domain. The idea of the SLM technique is to multiply the vector X by a vector $\Phi^{(u)} = \{\phi_k^{(u)}\}, k = 0, \dots, N - 1$. Where, $\phi_k^{(u)}$ are of the form:

$$\phi_k^{(u)} = e^{j\varphi_k^{(u)}}, \varphi_k^{(u)} \in [0, 2\pi), u = 0, \dots, U - 1 \quad (7)$$

The new OFDM signal in the frequency domain after weighting is written as: $X^{(u)} = X \cdot \Phi^{(u)}$. This gives different U signals of N components. Finally, the time transmitted OFDM signal is written:

$$X^{(u^*)} = IDFT(X^{(u^*)}) \quad (8)$$

Where $u^* = \arg \min_u \{\max_n |x_n^{(u)}|\}$ is the index corresponds to the OFDM signal having lowest PAPR. The value of the index u^* will then be transmitted to the receiver for the reconstruction via an error correcting code. In [36], the authors propose that the number of bits on which this index must be coded is of the order of $\log_2 U$. The principle of the technique is illustrated in Figure 3.

The SLM performs well in terms of reducing the PAPR. However, the major disadvantage of this technique is its complexity, due to the use of several (U) operations of IDFT. In addition, this method only requires the transmission of information sequences ("Side Information") so that the receiver identifies the sequence that allowed to generate the lowest PAPR. The disadvantage of transmitting information between the issuer and the receiver is twofold: first, because of the risk that this sequence is tainted with errors via the transmission and transmission channel.

G. Partial Transmit Sequences (PTS)

The PTS technique is proposed in [36] by S. Muller and J. B. Huber.

The idea of this method is to shorten the train of the N carriers in V blocks of $\frac{N}{V}$ carriers. A carrier used in a particular block will be set to zero in all others. Once these $\frac{N}{V}$ blocks are formed, the initial idea of SLM is applied: a vector $\Phi^{(v)} = \{\phi_k^{(v)}\}, v = 1, \dots, V$ will perform a weighting of each of the V blocks after IDFT to form the final signal to the reduced PAPR.

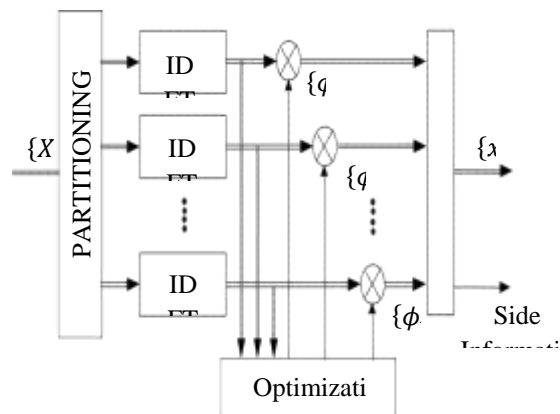


Figure 4: PTS scheme for PAPR reduction

As illustrated in Figures 4, the PTS algorithm is as follows:

The frequency OFDM symbol X of N carriers is shortened in V sub-blocks disjoint $X^{(v)}$ of $\frac{N}{V}$ carriers such as $X = \sum_{v=1}^V X^{(v)}$.

The frequency OFDM symbol X of N carriers is truncated in V sub-blocks disjoint FRED symbol OFDM is written:

$$X = \sum_{v=1}^V X^{(v)} \cdot \phi^{(v)}, \phi^{(v)} = e^{j\varphi^{(v)}} \quad v = 1, \dots, V \tag{9}$$

The temporal OFDM symbol x is then written:

$$x = IDFT \left(\sum_{v=1}^V X^{(v)} \cdot \phi^{(v)} \right) = \sum_{v=1}^V \phi^{(v)} \cdot IDFT(X^{(v)}) = \sum_{v=1}^V \phi^{(v)} \cdot x^{(v)} \tag{10}$$

Where the phase shift vector $\Phi^{(v)}$ is optimized in the following way:

$$\Phi^{(v)} = \{\phi^{(v)}\} \arg \min_{\phi^{(v)}} \left\{ \max_n \sum_{v=1}^V \phi^{(v)} \cdot x^{(v)} \right\} \tag{11}$$

$n = 0, 1, \dots, N - 1$

The way in which symbols are partitioned into sub-blocks has an influence on the intricacy of the method. The major disadvantage of the PTS technique lies in the complexity of finding the weight vectors ($\Phi^{(v)}$) to minimize the PAPR. Indeed, by considering V in blocks and binary weighting factors (the vectors $\Phi^{(v)}$, $v = 1, 2$ are only composed of 1 or -1), the number of possible combinations is 2^V which must all be reviewed to determine which vector set minimizes the PAPR. The idea proposed in [38] is then to stop the vector search process $\Phi^{(v)}$ when the desired PAPR is reached. Another disadvantage of the PTS technique is that it needs the Side Information (SI) in order for the receiver to identify the sequence that generated the lowest PAPR.

H. The Companding Technique (CT)

The Companding method is typically used in voice signals to optimize the number of bits per sample. Because the operation of voice signals and OFDM is analogous, in the sense that high peaks occur infrequently, The objective of the CTs technique is to convert the OFDM signal in the time domain depending on the power distribution of the signal to ensure that parts of the signal are attenuated with high peaks, while areas with low amplitudes are attenuated. Although the CTs scheme has a good performance in reduction of the PAPR, the BER increases because it introduces distortion in the transmitted signal; In addition, when trying to recover the original signal (decompander) on the receiver side, the channel noise could also be expanded [39].

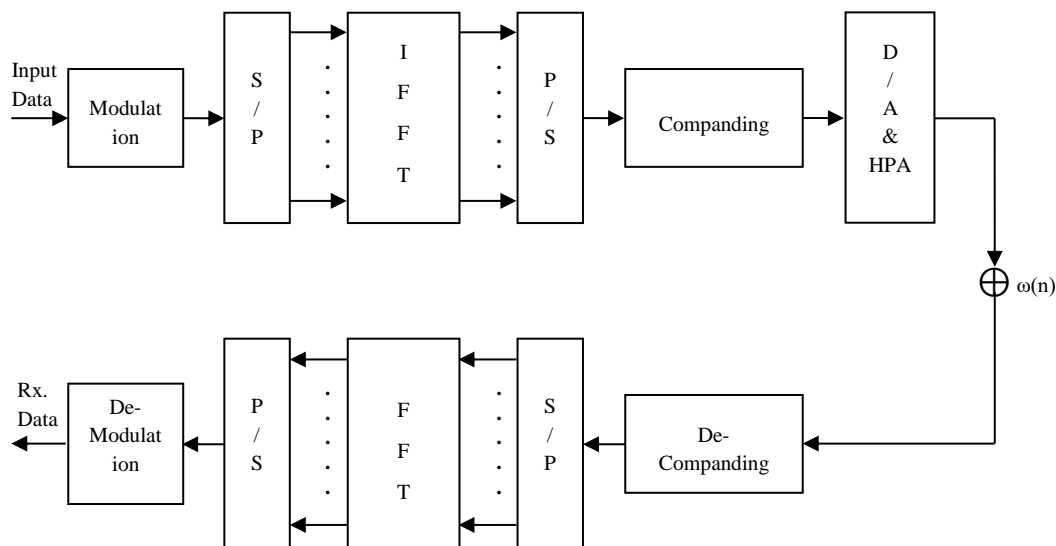


Figure 5: Companding scheme for PAPR reduction

The μ -law companding was one of the first non-linear techniques of PAPR reduction [40]. The technical μ -law CTs maintains the peaks of the signal, but increases the parts with lower amplitudes, thus obtaining a signal with a lower average power, which leads to a lower PAPR. At the receiver, the signal must be retrieved before demodulation. It is given by:

$$y = V \frac{\log[1+\mu\frac{|x|}{V}]}{\log(1+\mu)} \operatorname{sgn}(x) \quad (12)$$

Where V = peak amplitude of the signal.

x =amplitude of the input signal.

Decompression is achieved by applying the inverse of equation (12).

Here the value of μ is optimized using the Cuckoo Search algorithm which is explained in the following sub heading.

I. Cuckoo Search (CS)Algorithm

Cuckoo Search is a population-based search technique for the optimization of complex problems, non-linear and non-convex optimization.

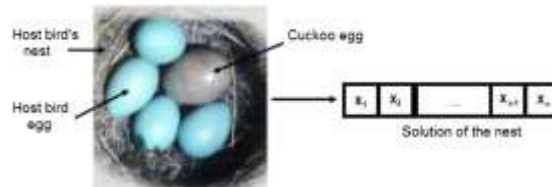


Figure 6: Representation of solutions in cuckoo search

Cuckoo species practice parasitism by depositing their eggs in the nests of another bird species. Host birds may come into conflict with their unwanted occupants (cuckoos) and, for example, if they discover that the eggs are not theirs, pull them from the nest or just leave the nest and fabricate another one in somewhere else. Some cuckoo species, for example, the Striped Cuckoo have evolved to such an extent that the colour and patterns of the eggs of these species are very similar to the eggs of the host species.

Rules of Cuckoo Search Algorithm:

- Each cuckoo deposits one egg at a time and leaves its egg in an arbitrarily chosen nest.
- The best nests with high egg quality are maintained for the next iteration.
- The number of nests existing in each generation is predefined, and the eggs will be discovered and discarded with a probability $p_\alpha \in [0,1]$. For ease a fraction of the total of the nests will be replaced by new nests each generation.
- In the Cuckoo Search algorithm, x_i is the i^{th} nid and $n \in N$ represents problem variables.

A recent study shows that fruit flies explore the environment using a series of straight flight paths with 90 degree turns, following what is called a Lévy flight pattern. This behaviour has been used in optimization and optimization search problems whose results have shown a promising capacity.

The implementation of the Levy flight in the Cuckoo Search algorithm is utilized to create a novel solution during the exploration process [41].

$$x_{t+\tau} = x_t + \Phi(t) \quad (13)$$

$$x_i^{t+1} = x_i^t + \alpha \oplus Levy(\lambda) \quad (14)$$

$$Levy = t^{-\lambda}, (1 < \lambda \leq 3) \quad (15)$$

Cuckoo Search Algorithm

1. Generate population of n host nests $x_i (i = 1, \dots, I)$;
2. Objective function $f(x), x = (x_1, x_2, \dots, x_n)^T$;
3. **for** $k \leftarrow 1$ **to** K **do**
4. Select a cuckoo (Say, i) at random by means of Levy flights;
5. Evaluate F_i quality / performance;
6. Calculate the value of the objective function $F(i)$;
7. Select a nest i between I (Say, j) randomly;
8. **if** $F_i > F_j$ **then**
9. Replace j as the new solution;
10. Leave a fraction p_α of the worst nests;
11. Build new nests in new locations through Levy flights;
12. Maintain the best solutions (nests with the best performance);
13. Classify the solutions and find the best current solution;
14. Display results;

SIMULATION AND RESULTS

MATLAB 14a has been used to simulate the proposed algorithms.

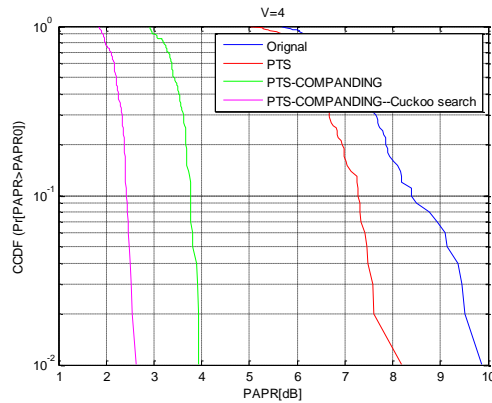


Figure 7: Comparative results of PTS Companding

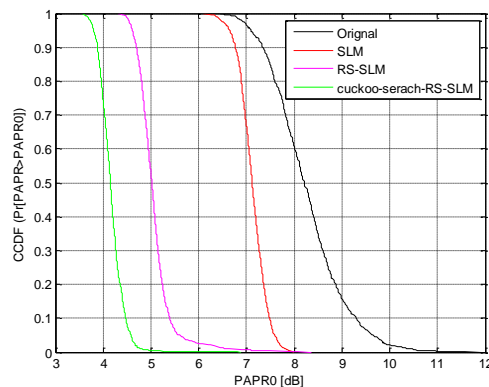


Figure 8: Comparative results for different techniques

CONCLUSION

This paper presented the importance of OFDM modulation in broadband communications systems and the appropriate treatment that should be given to the signal to reduce PAPR.

We have taken the opportunity to propose an extension of PAPR reduction techniques which is a hybrid of PTS Companding and further PAPR is improved with cuckoo search soft computing method.

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